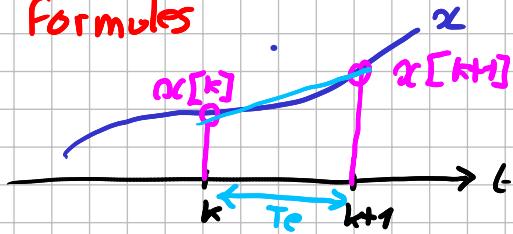
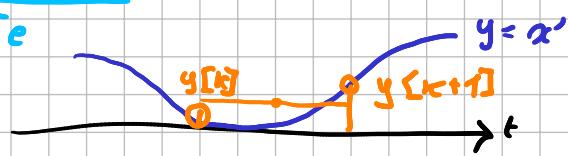


TD SYNTHèse Bilineaire

1-A Retrouver les Formules



$$1) \text{ pente} = \frac{x[k+1] - x[k]}{T_e}$$



$$2) \text{ moyenne} = \frac{y[k+1] + y[k]}{2} = \frac{x[k+1] - x[k]}{T_e}$$

$$3) z \cdot \frac{Y(z) + y(z)}{2} = \frac{X(z) - x(z)}{T_e}$$

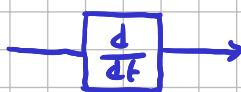
$$y(z) \cdot \frac{(z+1)}{2} = X(z) \cdot \frac{(z-1)}{T_e}$$

$$\frac{Y(z)}{X(z)} = \frac{2}{T_e} \frac{z-1}{z+1}$$

$$4) s(z) = \frac{2}{T_e} \frac{z-1}{z+1}$$

Équivalents harmoniques

p \leftrightarrow dérivateur $\frac{d}{dt}$ $e^{i\omega t}$
phaseur



$$(e^{i\omega t})' = i\omega \cdot e^{i\omega t}$$

phaseur de sortie

$$X(p) \xrightarrow{P} Y(p) = p \cdot X(p)$$

$$e^{i\omega t} \xrightarrow{j\omega} j\omega \cdot e^{i\omega t}$$

donc $p \equiv j\omega$ pour une onde pure
 p homogène à $[1/s] = [s^{-1}]$

z \leftrightarrow avance unitaire $e^{i\omega T_e}$

$$\frac{1}{z} \cdot e^{i\omega t} \xrightarrow{T^{-1}} e^{i\omega t}$$

$$= e^{i\omega(k+1)T_e} \cdot e^{i\omega k T_e}$$

phaseur de sortie

$$X(z) \xrightarrow{z} Y(z) = z \cdot X(z)$$

$$e^{i\omega T_e} \xrightarrow{e^{i\omega t}} e^{i\omega t} \cdot e^{i\omega T_e}$$

donc $z \equiv e^{i\omega T_e}$ pour une onde pure

z sans dimension

Vérifications

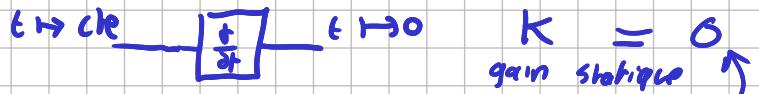
1) Homogénéité



$$\frac{y(v)}{x(v)} = \rho \quad \left[\frac{v/s}{v} \right] = \left[s^{-1} \right]$$

$$s(z) = \frac{y(z)}{x(z)} \left[\frac{v/s}{v} \right] = \frac{\frac{2}{Tc}}{s} \left[\frac{1+z}{1-z} \right] = \left[\frac{1+z}{s} \right] \quad \text{OK ✓}$$

2) Gain statique



$$\frac{y(\rho)}{x(\rho)} = \rho \equiv j\omega \xrightarrow[\substack{w \rightarrow 0 \\ \text{statique}}]{} 0 = K \quad \text{OK ✓}$$

$$s(z) = \frac{2}{Tc} \frac{1+z}{1-z}$$

$$z \equiv e^{j\omega Tc} \longrightarrow 1 \quad z=1 \equiv \text{statique en discret}$$

$$s(z) \equiv s(e^{j\omega Tc}) \xrightarrow[w \rightarrow 0]{} s(1) = \frac{2}{Tc} \frac{1+1}{1-1} = +\infty \quad \text{KO X}$$

Et oui ! c'est une erreur

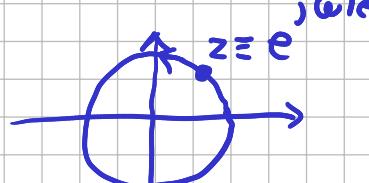
$$s(z) = \frac{2}{Tc} \frac{1-z}{1+z} = \frac{2}{Tc} \frac{1-1}{1+1} = 0 = K \quad \text{OK ✓}$$

3) Gain HF

$$\rho = j\omega \xrightarrow[\substack{w \rightarrow \infty}]{\substack{+j\infty}} \quad \rho = +j\infty \quad \text{pour le gain HF}$$



$$z \equiv e^{j\omega Tc}$$



$\omega \rightarrow \infty$ n'est pas HF en discret !

$$\omega = 2\pi \frac{f_c}{2} \rightarrow e^{j\omega Tc} \rightarrow \begin{matrix} +1 \\ -1 \end{matrix} \quad \text{c'est la HF du discret}$$

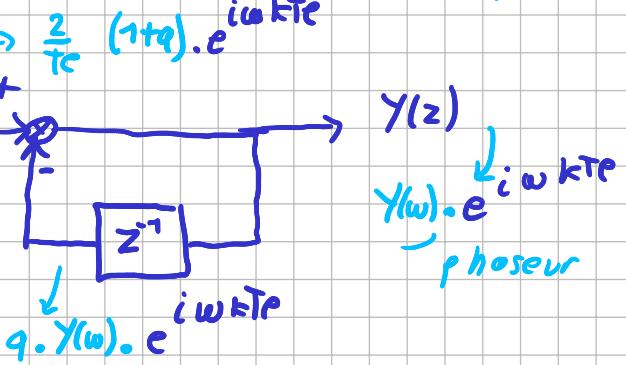
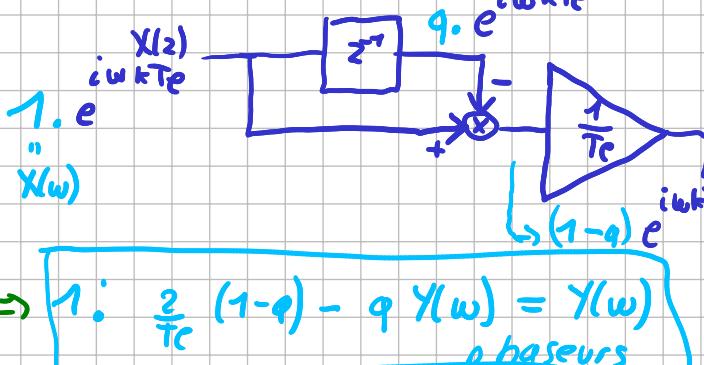
$$z \equiv e^{j2\pi f \cdot Tc} \xrightarrow[f \rightarrow \frac{f_c}{2}]{} e^{j\frac{2\pi}{2} f_c Tc} = e^{j\pi} = -1$$

En discret on fait $z = -1$ pour avoir le gain HF

$$s(z) \equiv s(e^{j\omega Tc}) \xrightarrow[w \rightarrow 2\pi \frac{f_c}{2}]{} s(-1) = \frac{2}{Tc} \frac{-1-1}{-1+1} = \infty \quad \text{OK ✓}$$

Équivalent harmonique de $s(z)$

$$z^{-1} \equiv e^{-j\omega Tc} \quad z \equiv e^{j\omega Tc} = \frac{1}{q}$$



$$\text{On résoud donc } \frac{2}{T_e} (1-q) \cdot X(\omega) = (1+q) Y(\omega)$$

$$s(\omega) = \frac{Y(\omega)}{X(\omega)} = \frac{2}{T_e} \frac{1+q}{1-q}$$

On ne le fera plus mais on voit que

$$s(z) = \frac{Y(z)}{X(z)} = \frac{2}{T_e} \frac{z-1}{z+1} = \frac{2}{T_e} \frac{1-2^{-1}}{1+2^{-1}} \Rightarrow \frac{Y(\omega)}{X(\omega)} = s(e^{i\omega T_e})$$

$$s(\omega) = \frac{Y(\omega)}{X(\omega)} = \frac{2}{T_e} \frac{q^{-1}}{q^{-1}+1} = \frac{2}{T_e} \frac{1-q}{1+q} \text{ avec } q = e^{-i\omega T_e} \stackrel{\text{III}}{=} \frac{1}{z^{-1}}$$

Calcul pénible :

$$s(\omega) = \frac{2}{T_e} \frac{1-e^{-i\omega T_e}}{1+e^{-i\omega T_e}} = \frac{2}{T_e} \underbrace{\frac{e^{-i\omega T_e/2}}{e^{i\omega T_e/2}}}_{i} \frac{\frac{e^{i\omega T_e/2}}{2} - e^{-i\omega T_e/2}}{e^{i\omega T_e/2} + e^{-i\omega T_e/2}}$$

$$s(\omega) = \frac{2}{T_e} \cdot \frac{2i \sin(\omega T_e/2)}{2 \cos(\omega T_e/2)} = i \frac{2}{T_e} \tan(\omega T_e/2)$$

$$s(z) \equiv s(\omega) = i \frac{2}{T_e} \tan(\omega \frac{T_e}{2}) \underset{w \rightarrow 0}{\sim} i \frac{2}{T_e} \omega \frac{T_e}{2} = i \omega$$

$$\text{En discret pour } \omega = \omega_d \quad s(z) \equiv i \frac{2}{T_e} \tan(\omega_d \frac{T_e}{2}) \underset{z \rightarrow \infty}{\sim} i \omega_d$$

$$\text{En continu pour } \omega = \omega_c \quad p \equiv j \omega_c$$

1-B Première Ordre.

$$0) \quad G(p) = \frac{k\tau}{1+\tau p} \text{ on veut } \frac{[V]}{[V]} = [-1] \quad p = j\omega \text{ donc } p [s^{-1}]$$

$$\text{on a } \frac{[1][s]}{1+[s][s^{-1}]} = [s] \quad \text{PB!} \times$$

$$\omega=0 \Rightarrow \text{On veut } G(p) \equiv G(j\omega) \xrightarrow{\omega \rightarrow 0} k \quad G(j\omega) = k \times \text{PB!}$$

$$\text{On corrige } G(p) = \frac{k}{1+\tau p}$$

$$\omega=\omega_c \Rightarrow \text{idem pour } f_c \text{ on veut } -3 \text{ dB}$$

$$\Rightarrow \frac{1}{1+\tau j\omega_c} = \frac{1}{1+j1} \text{ modulo } \frac{1}{\sqrt{2}} \quad \text{On corrige } \tau = \frac{1}{\omega_c} = \frac{1}{2\pi f_c}$$

$$G(p) = \frac{k}{1+\tau p}, \quad 2\pi\tau = \frac{1}{f_c}$$

$$1) \quad G_s(z) = G\left(\frac{2}{T_e} \frac{z-1}{z+1}\right) = \frac{k}{1+2\frac{2}{T_e} \frac{z-1}{z+1}} = k \frac{(1+z)}{1+z+2\frac{2}{T_e}(z-1)}$$

— issu de la correction

$$G_s(z) = k \frac{\frac{1}{z} + 1}{1 - 2\frac{T_c}{T_e} + (1 + 2\frac{T_c}{T_e})z}$$

$$G_s(1) = k \frac{2}{2} = k \quad \checkmark$$

$$G_s(-1) = k \frac{0}{-2\frac{T_c}{T_e}} = 0 \quad \checkmark$$

$G_s(z)$ [1] homogène \checkmark

$$G_s(z) = \frac{k}{1 + 2\frac{T_c}{T_e} z} = \frac{1 + z^{-1}}{1 + \left(\frac{T_c + 2T_e}{T_c - 2T_e}\right) z^{-1}}$$

$$G_s(1) = \frac{k \cdot 2}{1 + 2\frac{T_c}{T_e} + T_c - 2T_e} = k \text{ ouf!}$$

$$G_s(-1) = 0 \text{ ou } r$$

$$G_s(z) \text{ [1]} \quad \checkmark$$

$$G_s\left(\frac{1}{z}\right) = \frac{k \cdot T_c}{T_c + 2T_e} \frac{1 + z^{-1}}{1 + \left(\frac{T_c + 2T_e}{T_c - 2T_e}\right) z^{-1}}$$

$$G_s(1) = \frac{2k}{2} = k \quad \checkmark$$

$$G_s(-1) = k \frac{0}{4T_e} = 0 \quad \checkmark$$

$$G_s(z) \text{ [-1]} \quad \checkmark$$

$$G_s\left(\frac{1}{z}\right) = k' \frac{1 + q}{1 + a_1 q} \quad \text{avec}$$

$$k' = k \cdot \frac{T_c}{T_c + 2T_e} \quad \text{gain statique en discret}$$

$$a_1 = \frac{T_c - 2T_e}{T_c + 2T_e} \quad \text{[-1]}$$

$$5) \quad \frac{y(z)}{z} (1 + a_1 z^{-1}) = k' X(z) (1 + z^{-1})$$

$$y[k] + a_1 y[k-1] = k' (x[k] + x[k-1])$$

$$y[k] = -a_1 y[k-1] + k' (\underbrace{x[k] + x[k-1]}_{\text{Auto Regressive}} \underbrace{(x[k] + x[k-1])}_{\text{Moving Average}}) \quad \text{homogène}$$

6) Théorème suite extrait "si $(U_n)_{n \in \mathbb{N}} \rightarrow l$ " donc si stable ici $(y_n)_{n \in \mathbb{N}} \rightarrow y^*$ et $(y_{n+1}) \rightarrow y^*$ étro. en statique ($x_n = x^*$)

$$y^* = -a_1 y^* + k' (x^* + x^*) \Rightarrow y^* = \frac{2k'}{1 + a_1} x^*$$

$$y^* = \frac{2 \cdot \frac{K T_c}{T_c + 2T_e}}{1 + \frac{T_c - 2T_e}{T_c + 2T_e}} x^* = \frac{2k T_c}{T_c + 2T_e + T_c - 2T_e} = 2k \frac{T_c}{2T_e} = 2 \frac{k}{2} = k \quad \text{gain statique}$$

7) On sait qu' aux HF $\vec{w}_{F_n} = (1, -1, 1, -1) \Rightarrow x[k] = -x[k-1] = \dots = x^*$
en entrée $x^* \cdot \vec{w}_{F_n} \rightarrow \boxed{H} \underbrace{a_1 H(\frac{x^*}{z}) \cdot \vec{w}_{F_n}}_{y^*}$

En sortie on a aussi du $y^* \cdot \vec{w}_{F_n} \Rightarrow y[k] = -y[k-1] = \dots = y^*$
 $(y_{2n}) \rightarrow y^*$ et $(y_{2n+1}) \rightarrow -y^*$ si stable!

$$y[k+1] = \alpha_1 y[k] - k' (\alpha[k+1] + \alpha[k])$$

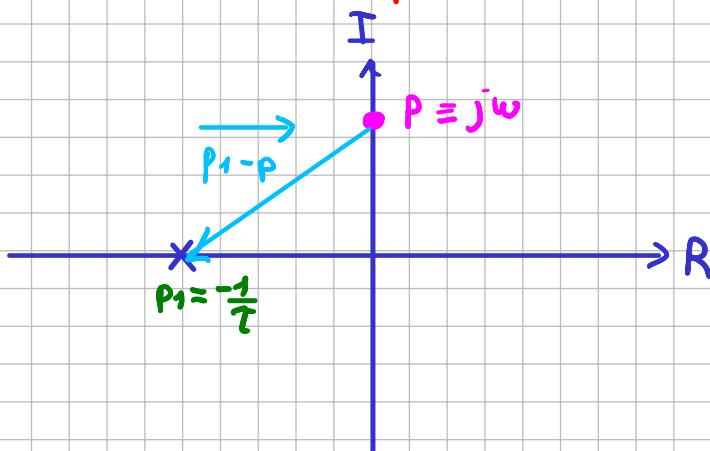
$\overrightarrow{w_{TF}}$

$$y^* = \alpha_1 \cdot -y^* - k' (-x^* + -x^*)$$

$$y^* = -\frac{k'}{1+\alpha_1} \cdot 0 = 0$$

gain HF du passe-bas

Réponse harmonique



$$\angle G(p) = \underbrace{\angle k}_{0 \text{ ou } \pi} + \underbrace{\angle p_1}_{\pi \text{ ou } 0} - \underbrace{\angle p_1 - p}_{\pi \text{ ou } 0}$$

$$G(p) = \frac{K}{1 + T_p p} = K \cdot \frac{1}{1 - \frac{p_1}{p - p_1}} = K \cdot \frac{p_1}{p - p_1}$$

$$p_1 = -\frac{1}{T} \quad \text{pôle } x$$

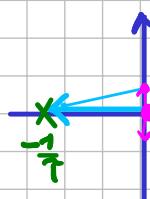
$$p = j\omega$$

$$|G(p)| = |K| \cdot \left| \frac{p_1}{p_1 - p} \right| \quad \text{dB} \quad |G(p)|_{\text{dB}} = |K|_{\text{dB}} + |p_1|_{\text{dB}} - |\overrightarrow{p_1-p}|_{\text{dB}}$$

Asymptotique

$$p = j\omega \xrightarrow{\omega \rightarrow \infty} j0$$

statique



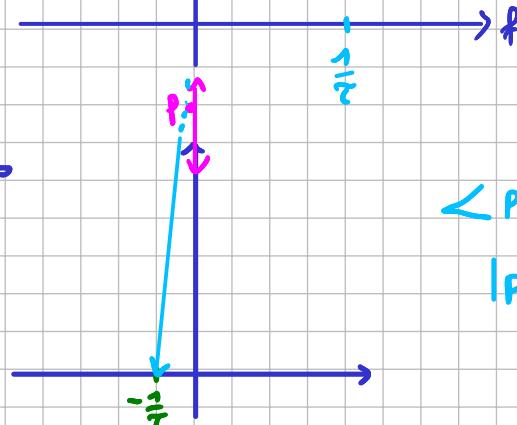
$$\left| \frac{1}{1 + T_p p} \right| \approx \frac{1}{1 + T_j \omega} \approx \frac{1}{\omega} \quad \text{si } \omega T \ll 1$$

$$\angle p_1 - p = \pi \Rightarrow \angle \text{pôle} = \angle p_1 - \angle p_1 - p = \pi \rightarrow \pi = 0$$

$$\angle G(p) \leq \left| \frac{1}{1 + T_p p} \right| \approx \frac{1}{1 + T_j \omega} \approx \frac{1}{\omega} \quad \text{si } \omega \ll \frac{1}{T}$$

$$\frac{K}{1 + T_p} \approx K \quad \omega \ll T$$

$$[HF] \quad p = j\omega \xrightarrow{\omega \rightarrow \infty} +j\infty$$



$$\angle p_1 - p = -\frac{\pi}{2} \Rightarrow \angle \text{pôle} = \pi - \frac{\pi}{2} = \frac{\pi}{2}$$

$$|p_1 - p| \approx |p| = |j\omega| = \omega \quad \frac{1}{\omega} \ll j\omega$$

$$\Rightarrow |\text{pôle}| \propto \frac{1}{\omega}$$

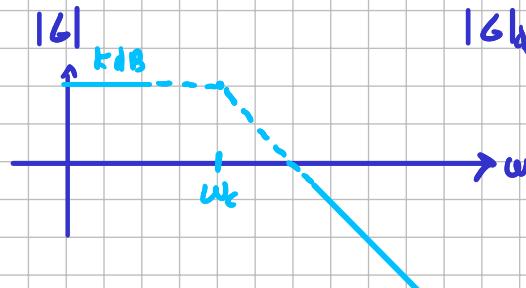
$$\frac{K}{1 + T_p} = \frac{K}{1 + T_j \omega} \approx \frac{K}{T_j \omega} \quad T_j \ll T_w \quad \frac{K}{T_j \omega} = -j \frac{K}{T_w} = K \cdot \left(-j \frac{\omega_e}{\omega} \right)$$

$$|\text{pôle}|_{\text{dB}} = \left| \frac{p_1}{p} + \frac{1}{\omega} \right|_{\text{dB}} = -20 \text{ dB/dec}$$

$$\omega \times 10 \Rightarrow \frac{|\text{pôle}|}{10}$$

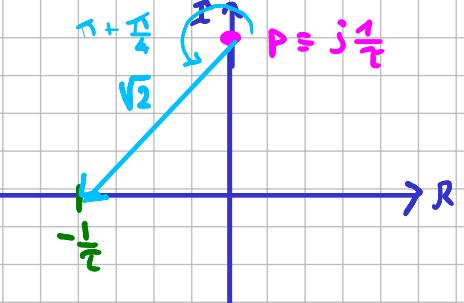
$$\angle G = -\pi/2$$

$$|G|_{\text{dB}} = K_{\text{dB}} - \left| \frac{\omega}{\omega_e} \right|_{\text{dB}}$$



Cassure

$$\omega = \omega_c$$



$$|pôle| = \sqrt{2}$$

$$\angle pôle = \pi - (\pi + \frac{\pi}{4}) = -\frac{\pi}{4}$$

$$\omega = \omega_c \Rightarrow G(j\omega) = \frac{k}{1 + j\omega \cdot \tau \cdot \omega_c} = \frac{k}{1 + j} \quad \omega_c = \frac{1}{\tau}$$



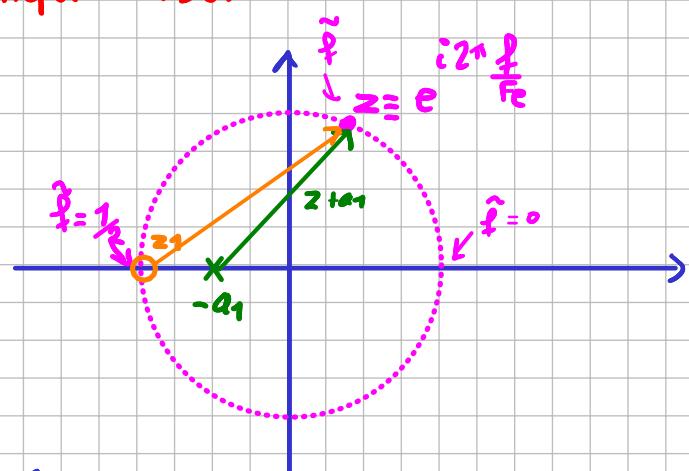
$$\angle G = \angle k - \angle 1+j \\ = 0 - \frac{\pi}{4}$$



Reponse harmonique discret

$$jTcw \rightarrow j2\pi f$$

$$z \equiv e^{j\omega} = e^{j2\pi f T}$$



$$G_s(1/q) = k' \frac{1+q}{1+a_1 q}$$

$$= k' \frac{z+1}{z+a_1}$$

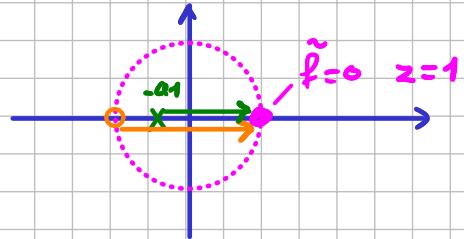
$$p_1 = -a_1$$

$$z_1 = -1$$

$$= k' \cdot -1 \frac{z - (-1)}{z - (-a_1)}$$

Statique

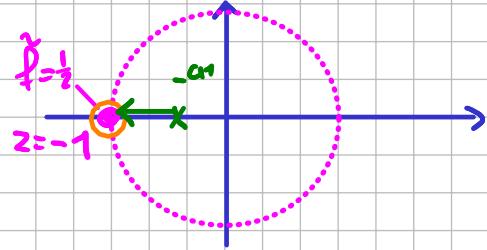
$$z \equiv e^{-j2\pi f \tilde{f}} \xrightarrow{\tilde{f}=0} 1 \quad G_s(z) = k' \frac{1+1}{1+a_1}$$



$$|G_s| = |k'| \frac{2}{a_1 + 1} = \frac{2k'}{a_1 + 1} = K$$

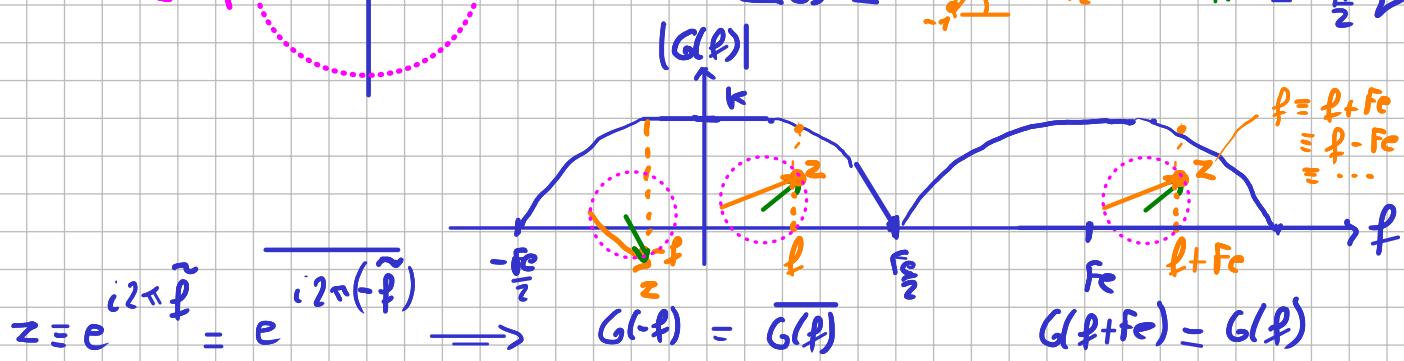
$$\angle G_s = 0^\circ - 0^\circ = 0^\circ \quad \checkmark$$

HF

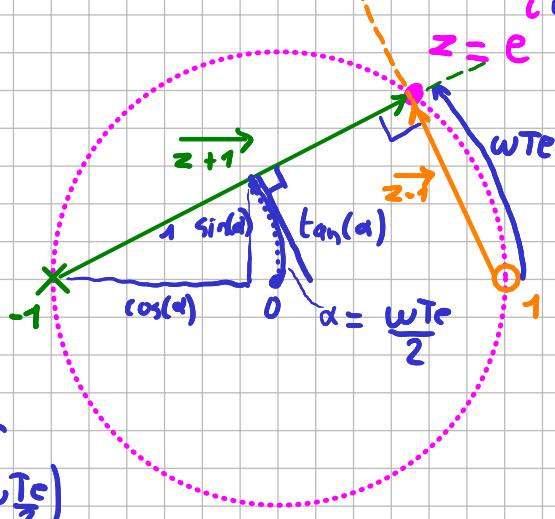


$$|G_s| = |k| \cdot \frac{0}{a_1} = 0 \quad \checkmark$$

$$\angle G_s = \pi/2 - \pi = -\pi/2 \quad \checkmark$$



$$\text{Retour sur } s(z) = \frac{2}{Te} \cdot \frac{z-1}{z+1} = \frac{2}{Te} \cdot \frac{z-(-1)}{z-(-1)}$$



$$|s(z)| = \frac{2}{T_e} \frac{|z-1|}{|z+1|} = \tan \alpha$$

$$= \tan \left(\omega \frac{T_e}{2} z \right)$$

$$\angle s(z) \equiv \angle z-1 - \frac{\pi}{2} + \alpha$$

$$\angle s(z) \equiv \pi \Rightarrow s(z) \equiv i \dots$$

$$s(z) \equiv j \frac{2}{T_e} \tan\left(\omega \frac{T_e}{2}\right)$$

Pour w_c

$$s(z) \equiv j \underbrace{\frac{2}{T_0} \tan(\omega_c \frac{T_0}{2})}_{\omega_d} = j \omega_d \text{ "pour discret"} \neq j \omega_c$$

$$\text{Dont} \quad G(s(z)) = G(j\omega) \neq G(j\omega) = \frac{k}{1+j\omega}$$

Prewarp p : On prend $G_{pre} = \frac{k}{1 + \tau_{pre} p}$ tel que $G_{pre}(j\omega) = \frac{k}{1 + j\omega}$

$$\text{Ainsi } G_{\text{pre}}(s(z)) \underset{\omega \rightarrow \omega_c}{\equiv} G_{\text{pre}}\left(j\omega_d = j\underbrace{\frac{2}{T_c} \tan(\omega_c \frac{T_c}{2})}_{\omega_d}\right) = \frac{k}{T+i}$$

$$\text{Done} \quad T_{\text{pre}} = \frac{1}{\omega_d} = \frac{1}{\frac{2}{T_E} \tan(\omega_c \frac{T_E}{2})}$$

si $\text{Fe} \nearrow$, $\text{Te} \searrow$ $w_c T_{\frac{g}{2}} \rightarrow$ et $\ll 1 \Rightarrow$ $w_d \underset{w_c}{\sim}$ w_c
prewarp inutile

$$G(z) = G_{\text{pre}}(s(z)) = k' \frac{1+z^{-1}}{1+\alpha_1 z^{-1}}$$

$$k' = k \frac{T_c}{T_c + 2T_{pre}}$$

$$T_{pre} = \frac{1}{\omega_d} \quad \text{car} \quad \omega_d = \frac{2}{T_p} \tan\left(\omega_c \frac{T_p}{2}\right)$$

$$\alpha_1 = \frac{T_c - 2T_{pre}}{T - 2T_{pre}}$$

$$G(\omega \rightarrow \infty) = k, \quad G(\omega \rightarrow \frac{\omega_0}{2}) = 0, \quad G(\omega \rightarrow \omega_c) = \frac{k}{\frac{1}{\omega_c} + i}$$